







ISO6731 SLASEY9B - DECEMBER 2019 - REVISED FEBRUARY 2023

ISO6731 General-Purpose Triple-Channel Digital Isolator with Robust EMC

1 Features

- 50 Mbps data rate
- Robust isolation barrier:
 - High lifetime at 1500 V_{RMS} working voltage
 - Up to 5000 V_{RMS} isolation rating
 - Up to 10 kV surge capability
 - ±150 kV/µs typical CMTI
- Wide supply range: 1.71 V to 1.89 V and 2.25 V to 5.5 V
- 1.71 V to 5.5 V level translation
- Default output high (ISO6731) and low (ISO6731F) options
- Wide temperature range: -40°C to 125°C
- 1.6 mA per channel typical at 1 Mbps
- Low propagation delay: 11 ns typical
- Robust electromagnetic compatibility (EMC)
 - System-level ESD, EFT, and surge immunity
 - ±8 kV IEC 61000-4-2 contact discharge protection across isolation barrier
 - Low emissions
- Wide-SOIC (DW-16) Package
- Safety-Related Certifications:
 - DIN EN IEC 60747-17 (VDE 0884-17)
 - UL 1577 component recognition program
 - IEC 62368-1, IEC 61010-1, IEC 60601-1 and GB 4943.1 certifications

2 Applications

- Power supplies
- Grid, Electricity meter
- Motor drives
- Factory automation
- **Building automation**
- Lighting
- **Appliances**

3 Description

The ISO6731 device is a high-performance, triplechannel digital isolators ideal for cost-sensitive applications requiring up to 5000 V_{RMS} isolation ratings per UL 1577. This device is also certified by VDE, TUV, CSA, and CQC.

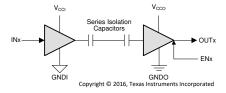
The ISO6731 devics provides high electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic input and output buffer separated by TI's double capacitive silicon dioxide (SiO₂) insulation barrier. This device comes with enable pins which can be used to put the respective outputs in high impedance for multi-master driving applications. The ISO6731 device has two forward and one reverse-direction channels. In the event of input power or signal loss, the default output is high for the device without suffix F and low for the device with suffix F. See Device Functional Modes section for further details.

Used in conjunction with isolated power supplies, this device helps prevent noise currents on data buses, such as UART, SPI, RS-485, RS-232, and CAN from damaging sensitive circuitry. Through innovative chip design and layout techniques, the electromagnetic compatibility of the ISO6731 device has been significantly enhanced to ease system-level ESD, EFT, surge, and emissions compliance. The ISO6731 device is available in a 16-pin SOIC widebody (DW) package and is a pin-to-pin upgrade to the older generations.

Device Information

PART NUMBER (1)	PACKAGE	BODY SIZE (NOM)
ISO6731, ISO6731F	` '	10.30 mm × 7.50
		mm

For all available packages, see the orderable addendum at the end of the data sheet.



V_{CCI}=Input supply, V_{CCO}=Output supply GNDI=Input ground, GNDO=Output ground

Simplified Schematic



Table of Contents

1 Features1	6.19 Switching Characteristics—1.8-V Supply	16
2 Applications1	6.20 Insulation Characteristics Curves	. 17
3 Description1	6.21 Typical Characteristics	.18
4 Revision History2	7 Parameter Measurement Information	. 19
5 Pin Configuration and Functions3	8 Detailed Description	.21
6 Specifications 4	8.1 Overview	.21
6.1 Absolute Maximum Ratings4	8.2 Functional Block Diagram	.21
6.2 ESD Ratings4	8.3 Feature Description	22
6.3 Recommended Operating Conditions5	8.4 Device Functional Modes	23
6.4 Thermal Information6	9 Application and Implementation	. 24
6.5 Power Ratings6	9.1 Application Information	. 24
6.6 Insulation Specifications7	9.2 Typical Application	. 24
6.7 Safety-Related Certifications8	10 Power Supply Recommendations	27
6.8 Safety Limiting Values8	11 Layout	. 28
6.9 Electrical Characteristics—5-V Supply9	11.1 Layout Guidelines	. 28
6.10 Supply Current Characteristics—5-V Supply9	11.2 Layout Example	. 29
6.11 Electrical Characteristics—3.3-V Supply10	12 Device and Documentation Support	30
6.12 Supply Current Characteristics—3.3-V Supply 10	12.1 Documentation Support	. 30
6.13 Electrical Characteristics—2.5-V Supply11	12.2 Receiving Notification of Documentation Updates.	.30
6.14 Supply Current Characteristics—2.5-V Supply 11	12.3 Support Resources	. 30
Electrical Characteristics—1.8-V Supply12	12.4 Trademarks	.30
6.15 Supply Current Characteristics—1.8-V Supply12	12.5 Electrostatic Discharge Caution	.30
6.16 Switching Characteristics—5-V Supply13	12.6 Glossary	
6.17 Switching Characteristics—3.3-V Supply14	13 Mechanical, Packaging, and Orderable	
6.18 Switching Characteristics—2.5-V Supply15	Information	. 30

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision A (March 2021) to Revision B (Febuary 2023)	Page
•	Changed standard name From: "DIN V VDE V 0884-11:2017-01" To: "DIN EN IEC 60747-17 (VDE 0884 throughout the document	-17)" 1
•	Updated CMTI From: "50 kV/us (min) and 75 kV/us (typ)" To: "100 kV/us (min) and 150 kV/us (typ)" throughout the document	1
•	Removed references to standard IEC/EN/CSA 60950-1 throughout the document	
	Removed standard revision and year references as well as "certification pending" references from all standard names throughout the document	1
•		
•	Updated V _{IORM} From: "1500 V _{PK} " To: "2121 V _{PK} " to match VDE certificate	
•	Updated V_{IOWM} From: "1060 V_{RMS} " To: "1500 V_{RMS} " and From: "1500 V_{DC} " To: "2121 V_{DC} " throughout the document.	
•	Added Maximum impulse voltage (V _{IMP}) specification in Insulation Specifications per DIN EN IEC 60747 (VDE 0884-17)	_
	Clarified method b test conditions of Apparent charge (q _{PD}) in Insulation Specification	
•)
•	Changed Figure 9-8 per DIN EN IEC 60747-17 (VDE 0884-17)	
C	hanges from Revision * (January 2021) to Revision A (March 2021)	Page
•	RTM release	1

5 Pin Configuration and Functions

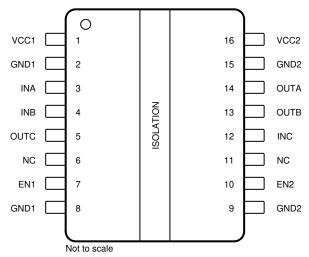


Figure 5-1. ISO6731 DW Package 16-Pin SOIC-WB Top View

Table 5-1. Pin Functions

	PIN	1/0	DESCRIPTION		
NAME ISO6731		I/O	DESCRIPTION		
EN1	7	I	Output enable 1. Output pins on side 1 are enabled when EN1 is high or open and in high-impedance state when EN1 is low.		
EN2	10	I	Output enable 2. Output pins on side 2 are enabled when EN2 is high or open and in high-impedance state when EN2 is low.		
GND1	2, 8	_	Ground connection for V _{CC1}		
GND2	9,15	_	Ground connection for V _{CC2}		
INA	3	I	Input, channel A		
INB	4	I	Input, channel B		
INC	12	I	Input, channel C		
NC	6,11		Not connected		
OUTA	14	0	Output, channel A		
OUTB	13	0	Output, channel B		
OUTC	5	0	Output, channel C		
V _{CC1}	1	_	Power supply, side 1		
V _{CC2}	16	_	Power supply, side 2		



6 Specifications

6.1 Absolute Maximum Ratings

See⁽¹⁾

		MIN	MAX	UNIT
Supply voltage (2)	V _{CC1} to GND1	-0.5	6	V
Supply voltage V	V _{CC2} to GND2	-0.5	6	
Input/Output	INx to GNDx	-0.5	V _{CCX} + 0.5 ⁽³⁾	V
Voltage	OUTx to GNDx	-0.5	V _{CCX} + 0.5 ⁽³⁾	
Output current	lo	-15	15	mA
Temperature	INx to GNDx OUTx to GNDx		150	°C
Temperature	Storage temperature, T _{stg}	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values
- (3) Maximum voltage must not exceed 6 V.

6.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001, all pins ⁽¹⁾	±6000	
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±1500	V
		Contact discharge per IEC 61000-4-2; Isolation barrier withstand test ⁽³⁾ (4)	±8000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
- (3) IEC ESD strike is applied across the barrier with all pins on each side tied together creating a two-terminal device.
- (4) Testing is carried out in air or oil to determine the intrinsic contact discharge capability of the device.

Submit Document Feedback

Copyright © 2023 Texas Instruments Incorporated

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V _{CC1} (1)	Supply Voltage Side 1	V _{CC} = 1.8 V	1.71		1.89	V
V _{CC1} (1)	Supply Voltage Side 1	V _{CC} = 2.5 V to 5 V	2.25		5.5	V
V _{CC2} (1)	Supply Voltage Side 2	V _{CC} = 1.8 V	1.71		1.89	V
V _{CC2} (1)	Supply Voltage Side 2	V _{CC} = 2.5 V to 5 V	2.25		5.5	V
Vcc (UVLO+)	UVLO threshold when supply	voltage is rising		1.53	1.71	V
Vcc (UVLO-)	UVLO threshold when supply	voltage is falling	1.1	1.41		V
Vhys (UVLO)	Supply voltage UVLO hystere	sis	0.08	0.13		V
V _{IH}	High level Input voltage		0.7 x V _{CCI}		V _{CCI}	V
V _{IL}	Low level Input voltage		0		0.3 x V _{CCI}	V
		V _{CCO} = 5 V ⁽²⁾	-4			mA
V _{IL}	Lligh level output ourrent	V _{CCO} = 3.3 V	-2			mA
ЮН	High level output current	V _{CCO} = 2.5 V	-1			mA
V _{IH} V _{IL} I _{OH}		V _{CCO} = 1.8 V	-1			mA
		V _{CCO} = 5 V			4	mA
	Low lovel output ourrest	V _{CCO} = 3.3 V			2	mA
l _{OL}	Low level output current	V _{CCO} = 2.5 V			1	mA
		V _{CCO} = 1.8 V			1	mA
DR	Data Rate	·	0		50	Mbps
T _A	Ambient temperature		-40	25	125	°C

 $[\]begin{array}{ll} \text{(1)} & V_{CC1} \text{ and } V_{CC2} \text{ can be set independent of one another} \\ \text{(2)} & V_{CCI} = \text{Input-side } V_{CC}; \, V_{CCO} = \text{Output-side } V_{CC} \\ \end{array}$



6.4 Thermal Information

		ISO673x	
	THERMAL METRIC(1)	DW (SOIC)	UNIT
		16 PINS	
R _{0JA}	Junction-to-ambient thermal resistance	73	°C/W
R _{0JC(top)}	Junction-to-case (top) thermal resistance	36.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	40.4	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	17	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	39.9	°C/W
R _{0JC(bot)}	Junction-to-case (bottom) thermal resistance	_	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.5 Power Ratings

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
ISO6731	SO6731							
P _D	Maximum power dissipation (both sides)	V _{CC1} = V _{CC2} = 5.5 V, T _{.I} = 150°C, C _{.I} =			117.5	mW		
P _{D1}	Maximum power dissipation (side-1)	15 pF, Input a 25-MHz 50% duty cycle			47.7	mW		
P _{D2}	Maximum power dissipation (side-2)	square wave			69.8	mW		

Product Folder Links: /SO6731



6.6 Insulation Specifications

DADAMETER		TEST CONDITIONS	VALUE	
	PARAMETER	TEST CONDITIONS	DW-16	UNIT
CLR	External clearance ⁽¹⁾	Shortest terminal-to-terminal distance through air	>8	mm
CPG	External creepage ⁽¹⁾	Shortest terminal-to-terminal distance across the package surface	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	μm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>600	V
	Material group	According to IEC 60664-1	ı	
	0	Rated mains voltage ≤ 600 V _{RMS}	I-IV	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 1000 V _{RMS}	1-111	1
DIN EN	IEC 60747-17 (VDE 0884-17) (2)		"	
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	2121	V _{PK}
V _{IOWM}	Maximum working isolation voltage	AC voltage; Time dependent dielectric breakdown (TDDB) Test; See Figure 9-8	1500	V _{RMS}
1011111		DC voltage	2121	V _{DC}
V_{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60 s (qualification); V _{TEST} = 1.2 x V _{IOTM} , t= 1 s (100% production)	7071	V _{PK}
V _{IMP}	Maximum impulse voltage ⁽³⁾	Tested in air, 1.2/50-µs waveform per IEC 62368-1	7692	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ⁽⁴⁾	V _{IOSM} ≥ 1.3 × V _{IMP} ; tested in oil (qualification test), 1.2/50-µs waveform per IEC 62368-1	10000	V _{PK}
		Method a, After Input-output safety test subgroup 2/3, $V_{ini} = V_{IOTM}$, $t_{ini} = 60$ s; $V_{pd(m)} = 1.2 \times V_{IORM}$, $t_m = 10$ s	≤5	
q _{pd}	Apparent charge ⁽⁵⁾	Method a, After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$, $t_{ini} = 60 \text{ s}$; $V_{pd(m)} = 1.6 \text{ x } V_{IORM}$, $t_m = 10 \text{ s}$	≤5	pC
		Method b: At routine test (100% production) and preconditioning (type test); V_{ini} = 1.2 x V_{IOTM} , t_{ini} = 1 s; $V_{pd(m)}$ = 1.875 x V_{IORM} , t_m = 1 s (method b1) or $V_{pd(m)}$ = V_{ini} , t_m = t_{ini} (method b2)	≤5	
C _{IO}	Barrier capacitance, input to output ⁽⁶⁾	V _{IO} = 0.4 x sin (2πft), f = 1 MHz	~1	pF
		V _{IO} = 500 V, T _A = 25°C	>10 ¹²	
R_{IO}	Isolation resistance ⁽⁶⁾	$V_{IO} = 500 \text{ V}, 100^{\circ}\text{C} \le T_{A} \le 125^{\circ}\text{C}$	>10 ¹¹	Ω
		V _{IO} = 500 V at T _S = 150°C	>10 ⁹	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577				
V _{ISO}	Maximum withstanding isolation voltage	$V_{TEST} = V_{ISO}$, $t = 60$ s (qualification), $V_{TEST} = 1.2$ x V_{ISO} , $t = 1$ s (100% production)	5000	V _{RMS}

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed-circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *safe electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air to determine the surge immunity of the package.
- (4) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
- (5) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (6) All pins on each side of the barrier tied together creating a two-terminal device.

Copyright © 2023 Texas Instruments Incorporated

Submit Document Feedback



6.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN EN IEC 60747-17 (VDE 0884-17)	Certified according to IEC 62368-1, IEC 61010-1 and IEC 60601	Certified according to UL 1577 Component Recognition Program	Certified according to GB 4943.1	Certified according to EN 61010-1 and EN 62368-1
Maximum transient isolation voltage, 7071 V _{PK} ; Maximum repetitive peak isolation voltage, 2121 V _{PK} ; Maximum surge isolation voltage, 10000 V _{PK}	5000 V _{RMS} insulation per CSA 62368-1, IEC 62368-1, CSA 61010-1 and IEC 61010-1, 1000 V _{RMS} basic and 600 V _{RMS} reinforced working voltage (pollution degree 2, material group I); 5000 V _{RMS} insulation per CSA 60601-1 and IEC 60601-1, 2 MOPP for 250 V _{RMS}	Single protection, 5000 V _{RMS}	Reinforced insulation, Altitude ≤ 5000 m, Tropical Climate, 700 V _{RMS} maximum working voltage	5000 V _{RMS} reinforced insulation per EN 61010-1 and EN 62368-1 up to working voltage of 600 V _{RMS}
Certificate number: 40040142	Master contract number: 220991	File number: E181974	Certificate number: CQC21001304083	Client ID number: 077311

6.8 Safety Limiting Values

Safety limiting(1) intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DW-16	PACKAGE					
		R _{θJA} =73°C/W, V _I = 5.5 V, T _J = 150°C, T _A = 25°C See Figure 6-1			311.4	mA
	Cafaty input autout as aunaly aussant	R _{θJA} = 73°C/W, V _I = 3.6 V, T _J = 150°C, T _A = 25°C See Figure 6-1			475.7	··· Λ
Is	Safety input, output, or supply current	$R_{\theta JA} = 73^{\circ}\text{C/W}, V_{I} = 2.75 \text{ V}, T_{J} = 150^{\circ}\text{C}, \\ T_{A} = 25^{\circ}\text{C} \\ \text{See Figure 6-1}$			622	mA
		$R_{\theta JA} = 73^{\circ}C/W$, $V_{I} = 1.89$ V, $T_{J} = 150^{\circ}C$, $T_{A} = 25^{\circ}C$ See Figure 6-1			905.1	mA
Ps	Safety input, output, or total power	R _{θJA} = 73°C/W, T _J = 150°C, T _A = 25°C See Figure 6-2			1712.4	mW
T _S	Maximum safety temperature				150	°C

The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S should not be exceeded. These limits vary with the ambient temperature, TA.

The junction-to-air thermal resistance, $R_{\theta JA}$, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

Submit Document Feedback

Copyright © 2023 Texas Instruments Incorporated

 $T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device.

 $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature. $P_S = I_S \times V_I$, where V_I is the maximum input voltage.

6.9 Electrical Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -4 mA; See Figure 7-1	V _{CCO} - 0.4 (1)		V
V _{OL}	Low-level output voltage	I _{OL} = 4 mA; See Figure 7-1		0.4	V
V _{IT+(IN)}	Rising input switching threshold			0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}		V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}		V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx		10	μΑ
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10		μΑ
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx		28	μΑ
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-28		μA
СМТІ	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 7-1	100	150	kV/us
C _i	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 5 \text{ V}$		2.8	pF

- V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} Measured from input pin to same side ground.

6.10 Supply Current Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6731							
Supply current - DC signal ⁽²⁾	V _I = V _{CCI} ⁽¹⁾ (ISO6731); V _I = 0 V (ISO	16731 with F suffix)	I _{CC1}		1.9	2.8	
	V = V(CC V(1000701), V = 0 V (1000701 Willin Sullix)		I _{CC2}		2.2	3.5	
	V _I = 0 V (ISO6731); V _I = VCC _I (ISO6731 with F suffix)		I _{CC1}		4.1	5.8	
			I _{CC2}		3.5	5.3	
	1 M	1 Mbps	I _{CC1}		2.9	4.2	mA
		i ivibps	I _{CC2}		3.0	4.8	ША
Supply current - AC signal	All channels switching with square	10 Mbss	I _{CC1}		3.4	4.8	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		4.2	6.1	
		EO Mbas	I _{CC1}		6.1	7.9	
		50 Mbps	I _{CC2}		9.4	11.9	

- (1)
- V_{CCI} = Input-side V_{CC} Supply current valid for ENx = V_{CCx} and ENx = 0V
- (3) Supply current valid for ENx = V_{CCx}



6.11 Electrical Characteristics—3.3-V Supply

V_{CC1} = V_{CC2} = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -2mA; See Figure 7-1	V _{CCO} - 0.2 ⁽¹⁾			V
V _{OL}	Low-level output voltage	I _{OL} = 2mA; See Figure 7-1			0.2	V
V _{IT+(IN)}	Rising input switching threshold				0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}			V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}			V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx			10	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10			μA
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx			30	uA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30			uA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 7-1	100	150		kV/us
C _i	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 3.3 \text{ V}$		2.8		pF

- (1) V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC}
- (2) Measured from input pin to same side ground.

6.12 Supply Current Characteristics—3.3-V Supply

V_{CC1} = V_{CC2} = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	TEST CONDITIONS		MIN	TYP	MAX	UNIT
ISO6731					-		
PARAMETER ISO6731 Supply current - DC signal (2) Supply current - AC signal (3)	$V_I = V_{CCI}$ (1)(ISO6731); $V_I = 0$ V (ISO6731 with F suffix)		I _{CC1}		1.9	2.7	
	V = VCC \ \(\(\(\)(1300731\)\), \(\)\ = 0 \(\)\\\(\)(130	0731 Willi P Sullix)	I _{CC2}		2.2	3.4	
	V _I = 0 V (ISO6731); V _I = VCC _I (ISO6731 with F suffix)		I _{CC1}		4.0	5.8	
			I _{CC2}		3.5	5.3	
	4.00	1 Mbno	I _{CC1}		2.8	4.1	m A
		1 Mbps	I _{CC2}		3.0	4.7	mA
Supply current - AC signal	All channels switching with square	10 Mbns	I _{CC1}		3.2	4.6	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		3.8	5.7	
	5	50.14	I _{CC1}		5.1	6.8	
		50 Mbps	I _{CC2}		7.5	9.9	

- (1) V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V
- (3) Supply current valid for ENx = V_{CCx}

6.13 Electrical Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$ (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -1mA; See Figure 7-1	V _{CCO} - 0.1 ⁽¹⁾			V
V _{OL}	Low-level output voltage	I _{OL} = 1mA; See Figure 7-1			0.1	V
V _{IT+(IN)}	Rising input switching threshold				0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}			V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}			V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx			10	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10			μA
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx			30	uA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30			uA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 7-1	100	150		kV/us
Ci	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 2.5 \text{ V}$		2.8		pF

 $[\]begin{array}{ll} \text{(1)} & \text{V_{CCI} = Input-side V_{CC}; V_{CCO} = Output-side V_{CC}} \\ \text{(2)} & \text{Measured from input pin to same side ground.} \\ \end{array}$

6.14 Supply Current Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6731							
Supply current - DC signal (2)	V V (1)((\$06731): V 0 V ((\$0	06731 with E cuffix)	I _{CC1}		1.9	2.7	
	$V_I = V_{CCI}$ (1)(ISO6731); $V_I = 0$ V (ISO6731 with F suffix)		I _{CC2}		2.2	3.4	
	V _I = 0 V (ISO6731); V _I = VCC _I (ISO6731 with F suffix)		I _{CC1}		4.0	5.7	
			I _{CC2}		3.5	5.3	
		1 Mbno	I _{CC1}		2.8	4.1	
		1 Mbps	I _{CC2}		3.0	4.7	mA
Supply current - AC signal	All channels switching with square	10 Mbns	I _{CC1}		3.1	4.5	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		3.6	5.4	
	50	50 Mb = 5	I _{CC1}		4.5	6.2	
		50 Mbps	I _{CC2}		6.4	8.7	

 ⁽¹⁾ V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V

⁽³⁾ Supply current valid for ENx = V_{CCx}



Electrical Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -1mA; See Figure 7-1	V _{CCO} - 0.1 (1)		V
V _{OL}	Low-level output voltage	I _{OL} = 1mA; See Figure 7-1		0.1	V
V _{IT+(IN)}	Rising input switching threshold			0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}		V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}		V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx		10	μΑ
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10		μΑ
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx		30	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30		μA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 7-1	100	150	kV/us
C _i	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 1.8 \text{ V}$		2.8	pF

- V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} Measured from input pin to same side ground.

6.15 Supply Current Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6731							
Supply current - DC signal (2)	V V (1)(ISO6731): V 0 V (ISO	$V_1 = V_{CCI}$ (1)(ISO6731); $V_1 = 0$ V (ISO6731 with F suffix)			1.5	2.4	
			I _{CC2}		2	3.4	
	$V_1 = 0 \text{ V (ISO6731)}; V_1 = \text{VCC}_1 \text{ (ISO6731 with F suffix)}$		I _{CC1}		3.4	5.4	
			I _{CC2}		3.2	5.3	
	4 Mhm	1 Mbps	I _{CC1}		2.4	3.8	mA
			I _{CC2}		2.7	4.6	ША
Supply current - AC signal	All channels switching with square	10 Mbns	I _{CC1}		2.6	4.1	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		3.2	5.1	
	50	FO Mbno	I _{CC1}		3.7	5.3	
		50 Mbps	I _{CC2}		5.2	7.4	

- (1)
- V_{CCI} = Input-side V_{CC} Supply current valid for ENx = V_{CCx} and ENx = 0V
- (3) Supply current valid for ENx = V_{CCx}

Submit Document Feedback

Copyright © 2023 Texas Instruments Incorporated

6.16 Switching Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@ 100 kbps		11	18	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 7-1		0.2	7	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				6	ns
t _r	Output signal rise time	Con Figure 7.4		2.6	4.5	ns
t _f	Output signal fall time	See Figure 7-1		2.6	4.5	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output			18.6	25.8	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			18.6	25.8	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO673x	See Figure 7-2		14.2	21.1	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO673x			14.2	21.1	ns
t _{PU}	Time from UVLO to valid output data				300	μs
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2 V. See Figure 7-3		0.1	0.3	μs
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ t_{sk(o)} is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

6.17 Switching Characteristics—3.3-V Supply

V_{CC1} = V_{CC2} = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		11	18	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} – t _{PLH}	See Figure 7-1		0.5	7	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				7	ns
t _r	Output signal rise time	See Figure 7.1		1.6	3.2	ns
t _f	Output signal fall time	<u> </u>		1.6	3.2	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output	but		23.2	34.4	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			23.2	34.4	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO673x	See Figure 7-2		16.6	23	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO673x			16.6	23	ns
t _{PU}	Time from UVLO to valid output data				300	μs
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 7-3		0.1	0.3	μs
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

Submit Document Feedback

14

Product Folder Links: ISO6731

⁽²⁾ t_{sk(o)} is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

6.18 Switching Characteristics—2.5-V Supply

V_{CC1} = V_{CC2} = 2.5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		12	20.5	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 7-1		0.6	7.1	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				7	ns
t _r	Output signal rise time	Con Firmer 7.4		2	4	ns
t _f	Output signal fall time	See Figure 7-1		2	4	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output	put		28.1	43	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			28.1	43	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO673x			20.4	36.3	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO673x			20.4	36.3	ns
t _{PU}	Time from UVLO to valid output data				300	μs
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 7-3		0.1	0.3	μs
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ t_{sk(o)} is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



6.19 Switching Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN TY	P MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps	1	5 24	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} – t _{PLH}	See Figure 7-1	0	7 8.2	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels		6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾			8.8	ns
t _r	Output signal rise time	See Figure 7.4	2	.7 5.3	ns
t _f	Output signal fall time	<u> </u>	2	.7 5.3	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output	-	40	.3 63	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output		40	.3 63	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO673x	See Figure 7-2	3	51.4	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO673x		3	51.4	ns
t _{PU}	Time from UVLO to valid output data			300	μs
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 7-3	0	.1 0.3	μs
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1	ns

⁽¹⁾ Also known as pulse skew.

Product Folder Links: ISO6731

⁽²⁾ $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

6.20 Insulation Characteristics Curves

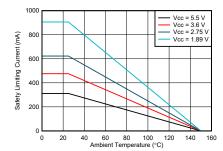


Figure 6-1. Thermal Derating Curve for Safety Limiting Current for DW-16 Package

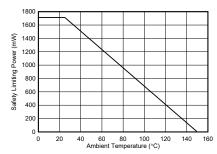


Figure 6-2. Thermal Derating Curve for Safety Limiting Power for DW-16 Package



6.21 Typical Characteristics

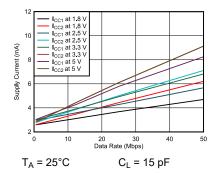


Figure 6-3. ISO6731 Supply Current vs Data Rate (With 15-pF Load)

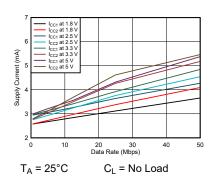
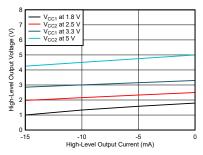
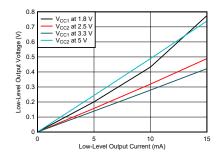


Figure 6-4. ISO6731 Supply Current vs Data Rate (With No Load)



 $T_A = 25$ °C



 $T_A = 25^{\circ}C$

Figure 6-5. High-Level Output Voltage vs High-level Figure 6-6. Low-Level Output Voltage vs Low-Level Output Current Output Current

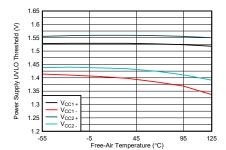


Figure 6-7. Power Supply Undervoltage Threshold vs Free-Air Temperature

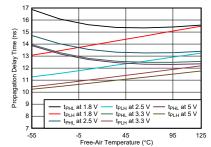
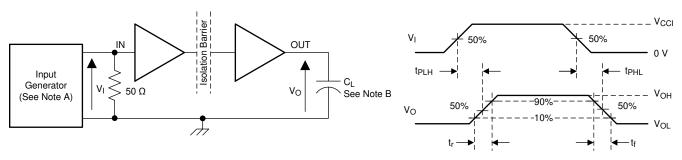


Figure 6-8. Propagation Delay Time vs Free-Air Temperature

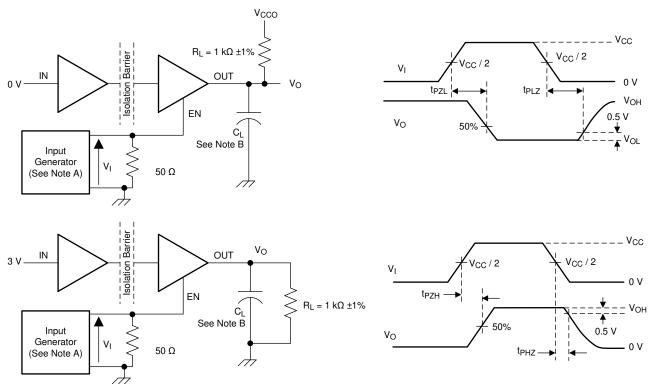
7 Parameter Measurement Information



Copyright © 2016, Texas Instruments Incorporated

- A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 50 kHz, 50% duty cycle, $t_r \leq$ 3 ns, $t_f \leq$ 3ns, $Z_O = 50 \Omega$. At the input, 50 Ω resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.

Figure 7-1. Switching Characteristics Test Circuit and Voltage Waveforms

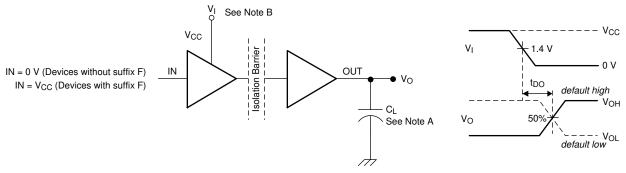


Copyright © 2016, Texas Instruments Incorporated

- A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 10 kHz, 50% duty cycle, $t_r \leq$ 3 ns, $t_f \leq$ 3 n
- B. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

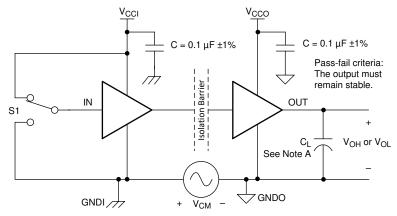
Figure 7-2. Enable/Disable Propagation Delay Time Test Circuit and Waveform





- A. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.
- B. Power Supply Ramp Rate = 10 mV/ns

Figure 7-3. Default Output Delay Time Test Circuit and Voltage Waveforms



A. $C_L = 15 \text{ pF}$ and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 7-4. Common-Mode Transient Immunity Test Circuit

Submit Document Feedback

Copyright © 2023 Texas Instruments Incorporated

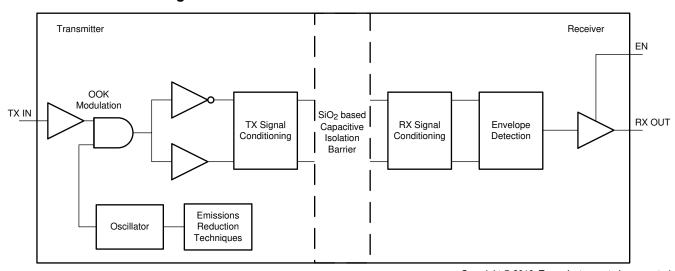


8 Detailed Description

8.1 Overview

The ISO6731 device has an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. If the ENx pin is low then the output goes to high impedance. The ISO6731 device also incorporate advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions due to the high frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, Figure 8-1, shows a functional block diagram of a typical channel.

8.2 Functional Block Diagram



Copyright © 2016, Texas Instruments Incorporated

Figure 8-1. Conceptual Block Diagram of a Digital Capacitive Isolator

Figure 8-2 shows a conceptual detail of how the ON-OFF keying scheme works.

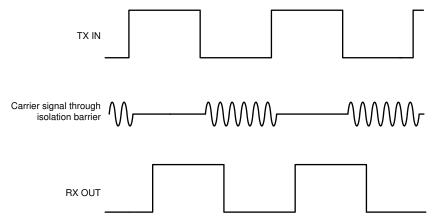


Figure 8-2. On-Off Keying (OOK) Based Modulation Scheme

Copyright © 2023 Texas Instruments Incorporated

Submit Document Feedback



8.3 Feature Description

Table 8-1 provides an overview of the device features.

Table 8-1. Device Features

PART NUMBER	CHANNEL DIRECTION	MAXIMUM DATA RATE	DEFAULT OUTPUT	PACKAGE	RATED ISOLATION(1)
ISO6731	2 Forward, 1 Reverse	50 Mbps	High	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6731F	2 Forward, 1 Reverse	50 Mbps	Low	DW-16	5000 V _{RMS} / 8000 V _{PK}

⁽¹⁾ See for detailed isolation ratings.

8.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 22. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO6731 device incorporates many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection cells for input and output signal pins and inter-chip bond pads.
- · Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- · Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.

Product Folder Links: ISO6731

8.4 Device Functional Modes

Table 8-2 lists the functional modes for the ISO6731 device.

Table 8-2. Function Table

	Table 6-2. FullCtion Table											
V _{CCI} ⁽¹⁾	V _{cco}	INPUT (INx) ⁽³⁾	OUTPUT ENABLE (ENx)	OUTPUT (OUTx)	COMMENTS							
		Н	H or open	Н	Normal Operation: A channel output assumes the logic state of its							
5	511	L	H or open	L	input.							
PU	PU	Open H or open		Default	Default mode: When INx is open, the corresponding channel output goes to its default logic state. Default is <i>High</i> for ISO6731 and <i>Low</i> for ISO6731 with F suffix.							
Х	PU	Х	L	Z	A low value of output enable causes the outputs to be high-impedance.							
PD	PU	Х	H or open	Default	Default mode: When $V_{\rm CCI}$ is unpowered, a channel output assumes the logic state based on the selected default option. Default is High for ISO6731 and Low for ISO6731 with F suffix. When $V_{\rm CCI}$ transitions from unpowered to powered-up, a channel output assumes the logic state of the input. When $V_{\rm CCI}$ transitions from powered-up to unpowered, channel output assumes the selected default state.							
Х	PD	х	x	Undetermined	When V_{CCO} is unpowered, a channel output is undetermined ⁽²⁾ . When V_{CCO} transitions from unpowered to powered-up, a channel output assumes the logic state of the input.							

- V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} ; PU = Powered up ($V_{CC} \ge 1.71$ V); PD = Powered down ($V_{CC} \le 1.05$ V); X = Irrelevant; H = High level; L = Low level; Z = High Impedance (1)
- The outputs are in undetermined state when 1.89 V < V_{CCI} , V_{CCO} < 2.25 V and 1.05 V < V_{CCI} , V_{CCO} < 1.71 V A strongly driven input signal can weakly power the floating V_{CC} through an internal protection diode and cause undetermined output

8.4.1 Device I/O Schematics

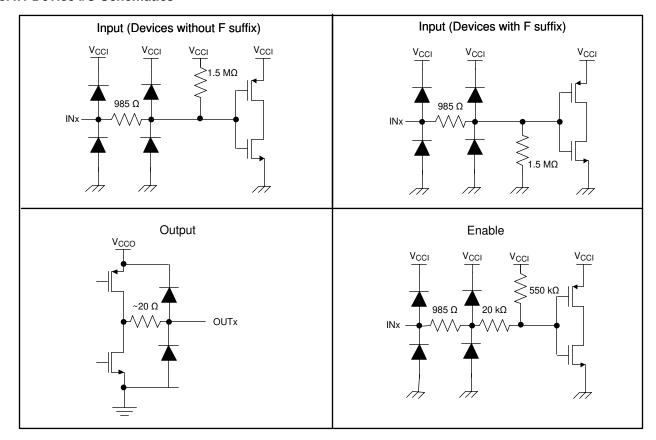


Figure 8-3. Device I/O Schematics

9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The ISO6731 device is a high-performance, triple-channel digital isolators. This device comes with enable pins on each side which can be used to put the respective outputs in high impedance for multi master driving applications. The ISO6731 device uses single-ended CMOS-logic switching technology. The supply voltage range is from 1.71 V to 5.5 V for both supplies, V_{CC1} and V_{CC2} . Since an isolation barrier separates the two sides, each side can be sourced independently with any voltage within recommended operating conditions. As an example, it is possible to supply ISO6731 V_{CC1} with 3.3 V (which is within 1.71 V to 5.5 V) and V_{CC2} with 5V (which is also within 1.71 V to 5.5 V). You can use the digital isolator as a logic-level translator in addition to providing isolation. When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, MCU or FPGA), and a data converter or a line transceiver, regardless of the interface type or standard.

9.2 Typical Application

Figure 9-1 shows the ISO6731 device, combined with Texas Instruments' mixed-signal microcontroller, RS-485 transceiver, transformer driver, and voltage regulator, can create an isolated RS-485 system.

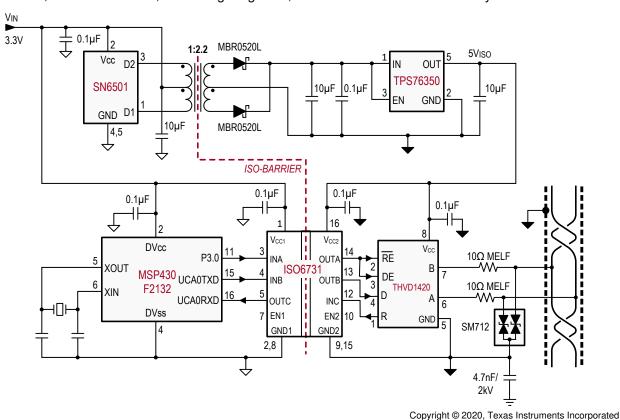


Figure 9-1. Isolated RS-485 Interface Circuit

Submit Document Feedback

Copyright © 2023 Texas Instruments Incorporated

9.2.1 Design Requirements

To design with this device, use the parameters listed in Table 9-1.

Table 9-1. Design Parameters

PARAMETER	VALUE
Supply voltage, V _{CC1} and V _{CC2}	1.71 V to 1.89 V and 2.25 V to 5.5 V
Decoupling capacitor between V _{CC1} and GND1	0.1 µF
Decoupling capacitor from V _{CC2} and GND2	0.1 µF

9.2.2 Detailed Design Procedure

Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO6731 device only requires two external bypass capacitors to operate.

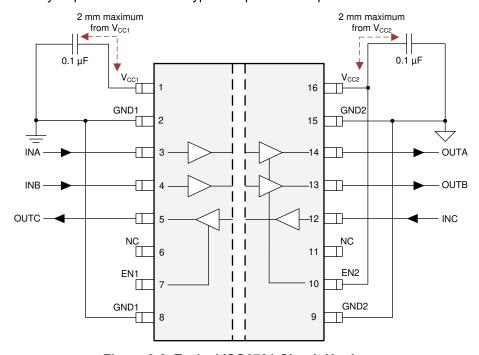
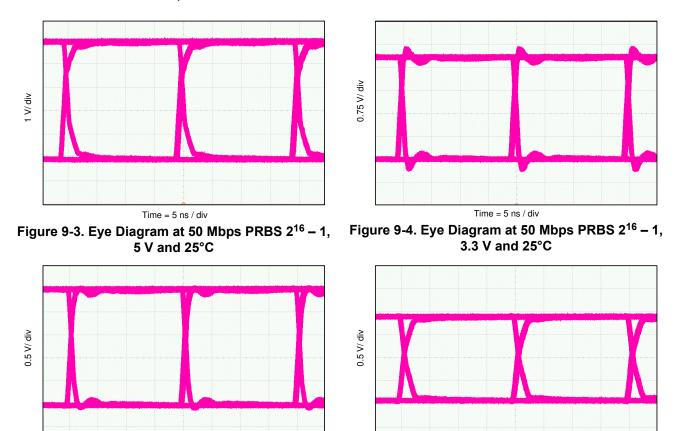


Figure 9-2. Typical ISO6731 Circuit Hook-up

9.2.3 Application Curve

The following typical eye diagrams of the ISO6731 family of devices indicates low jitter and wide open eye at the maximum data rate of 50 Mbps.



Time = 5 ns / div 2.5 V and 25°C

Time = 5 ns / div Figure 9-5. Eye Diagram at 50 Mbps PRBS 2¹⁶ – 1, Figure 9-6. Eye Diagram at 50 Mbps PRBS 2¹⁶ – 1, 1.8 V and 25°C

9.2.3.1 Insulation Lifetime

Insulation lifetime projection data is collected by using industry-standard Time Dependent Dielectric Breakdown (TDDB) test method. In this test, all pins on each side of the barrier are tied together creating a two-terminal device and high voltage applied between the two sides; See Figure 9-7 for TDDB test setup. The insulation breakdown data is collected at various high voltages switching at 60 Hz over temperature. For reinforced insulation, VDE standard requires the use of TDDB projection line with failure rate of less than 1 part per million (ppm). Even though the expected minimum insulation lifetime is 20 years at the specified working isolation voltage, VDE reinforced certification requires additional safety margin of 20% for working voltage and 50% for lifetime which translates into minimum required insulation lifetime of 30 years at a working voltage that's 20% higher than the specified value.

Figure 9-8 shows the intrinsic capability of the isolation barrier to withstand high voltage stress over its lifetime. Based on the TDDB data, the intrinsic capability of the insulation is 1500 V_{RMS} with a lifetime of 36 years. Other factors, such as package size, pollution degree, material group, etc. can further limit the working voltage of the component. The working voltage of DW-16 package is specified upto 1500 V_{RMS}. At the lower working voltages, the corresponding insulation lifetime is much longer than 36 years.

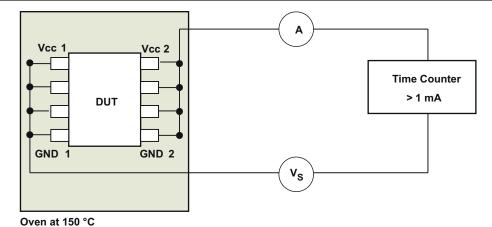


Figure 9-7. Test Setup for Insulation Lifetime Measurement

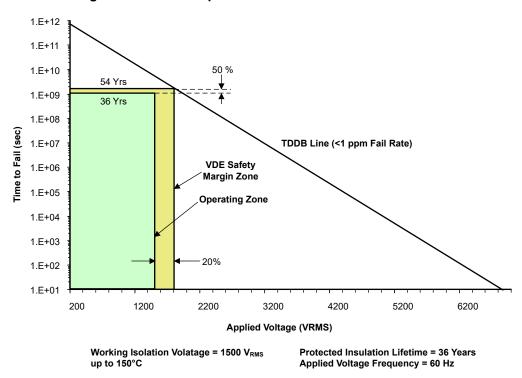


Figure 9-8. Insulation Lifetime Projection Data

10 Power Supply Recommendations

Power Supply Recommendation update with SN6505B (previously SN6505A)

To help ensure reliable operation at data rates and supply voltages, a 0.1- μ F bypass capacitor is recommended at the input and output supply pins (V_{CC1} and V_{CC2}). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver. For industrial applications, please use Texas Instruments' SN6501 or SN6505B. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501 Transformer Driver for Isolated Power Supplies or SN6505B-Q1 Low-noise, 1-A Transformer Drivers for Isolated Power Supplies



11 Layout

11.1 Layout Guidelines

A minimum of two layers is required to accomplish a cost optimized and low EMI PCB design. To further improve EMI, a four layer board can be used (see Figure 11-2). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch².
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the Digital Isolator Design Guide.

11.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit boards. This PCB is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.

Product Folder Links: ISO6731



11.2 Layout Example

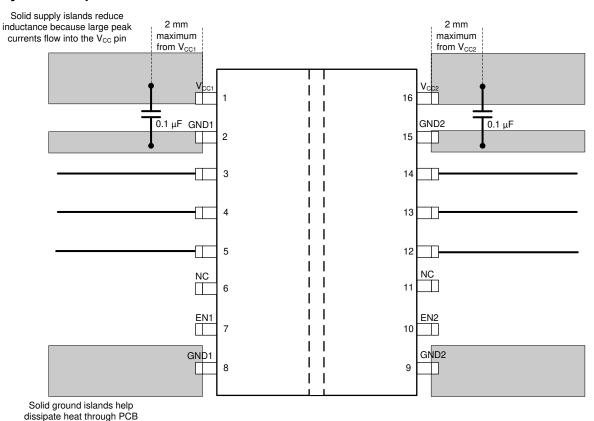


Figure 11-1. Layout Example

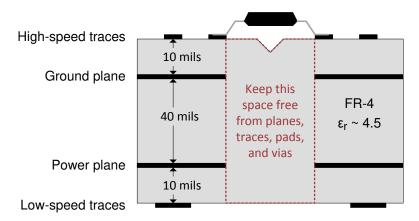


Figure 11-2. Layout Example Schematic



12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, Digital Isolator Design Guide
- Texas Instruments, ADS79xx 12/10/8-Bit, 1 MSPS, 16/12/8/4-Channel, Single-Ended, MicroPower, Serial Interface ADCs data sheet
- Texas Instruments, Isolation Glossary
- Texas Instruments, Top 6 Design Questions about I²C isolators
- Texas Instruments, Designing a reinforced isolated I²C-Bus interface by using digital isolators
- Texas Instruments, How to isolate signal and power for I²C interfaces
- Texas Instruments, How to use isolation to improve ESD, EFT, and Surge immunity in industrial systems
 application report
- Texas Instruments, DAC161P997 Single-Wire 16-bit DAC for 4- to 20-mA Loops data sheet
- Texas Instruments, MSP430G2132 Mixed Signal Microcontroller data sheet
- Texas Instruments, SN6501 Transformer Driver for Isolated Power Supplies data sheet
- Texas Instruments, TPS76333 Low-Power 150-mA Low-Dropout Linear Regulators data sheet

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

12.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.6 Glossary

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated device. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Product Folder Links: ISO6731

www.ti.com 26-Jul-2022

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
ISO6731DWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6731	Samples
ISO6731FDWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6731F	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE OPTION ADDENDUM

www.ti.com 26-Jul-2022

OTHER QUALIFIED VERSIONS OF ISO6731:

Automotive : ISO6731-Q1

NOTE: Qualified Version Definitions:

• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

www.ti.com 1-May-2024

TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

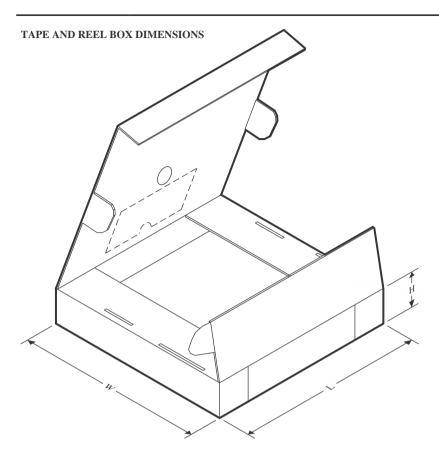
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO6731DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6731DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6731FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

www.ti.com 1-May-2024



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO6731DWR	SOIC	DW	16	2000	356.0	356.0	35.0
ISO6731DWR	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6731FDWR	SOIC	DW	16	2000	356.0	356.0	35.0

7.5 x 10.3, 1.27 mm pitch

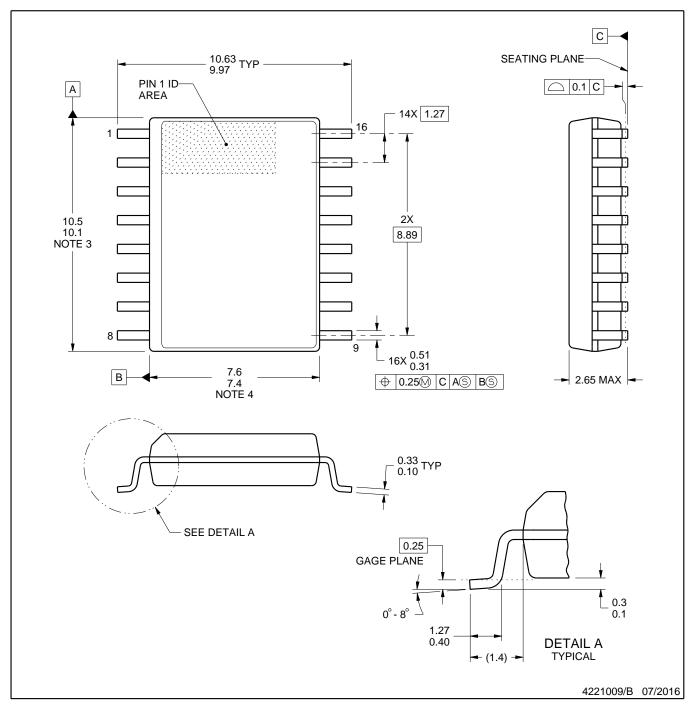
SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





SOIC



NOTES:

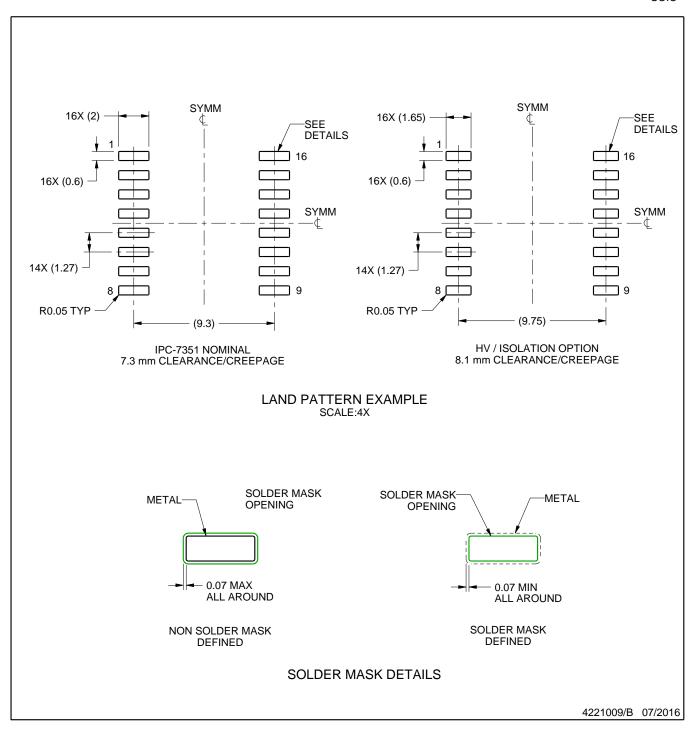
- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.



SOIC



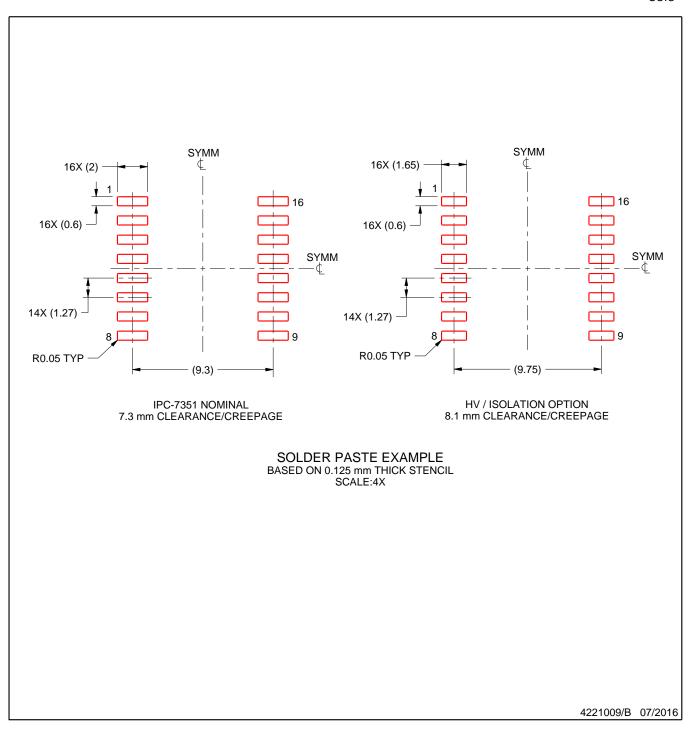
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOIC



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATA SHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

TI objects to and rejects any additional or different terms you may have proposed.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265 Copyright © 2024, Texas Instruments Incorporated